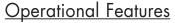


48Vin 12Vout 8.3A

High Efficiency, No Heatsink, Isolated DC/DC Converter

The PQ48120QNA08 PowerQorTM quarter brick converter is a next-generation, board-mountable, isolated, fixed switching frequency dc/dc converter that uses synchronous rectification to achieve extremely high conversion efficiency. The power dissipated by the converter is so low that a heatsink is not required, which saves cost, weight, height, and application effort. All of the power and control components are mounted to the multi-layer PCB substrate with high-yield surface mount technology. Since the PowerQor converter has no explicit thermal connections, it is extremely reliable.



- Ultra-high efficiency, 88% at full rated load current
- Delivers up to 100 Watts of output power with minimal derating - no heatsink required
- Wide input voltage range: 35V 75V meets or exceeds all 48V bus standards
- Fixed frequency switching provides predictable EMI performance
- No minimum load requirement means no preload resistors required

Mechanical Features

- Industry standard pin-out configuration (pin for pin compatible with Lucent QHW series)
- Industry standard size: 1.45" x 2.3"
- Total height less than 0.40", permits better airflow and smaller card pitch
- Total weight: 34 grams (1.2 oz.), lower mass greatly reduces vibration and shock problems

Safety Features

- 2000V, 10 $M\Omega$ input-to-output isolation provides input/output ground separation
- UL 1950 recognized (US & Canada), basic insulation rating
- TUV certified to EN60950

Power



PQ48120QNA08 Module

Protection Features

- Input under-voltage lockout disables converter at low input voltage conditions
- Output current limit and short circuit protection protects converter from excessive load current or short circuits
- Output over-voltage protection protects load from damaging voltages
- Thermal shutdown protects converter from abnormal environmental conditions

Control Features

- On/Off control referenced to input side (positive and negative logic options are available)
- Remote sense for the output voltage compensates for output distribution drops
- Output voltage trim: +10%/-20%, permits custom voltages and voltage margining

Safety Features (cont.)

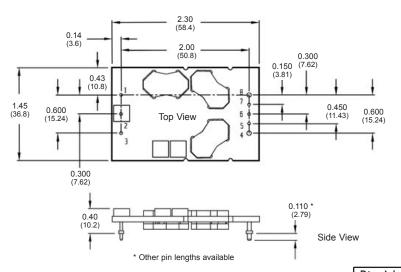
- Meets 72/23/EEC and 93/68/EEC directives which facilitates CE Marking in user's end product
- Board and plastic components meet 94V-0 flammability requirements

Product # PQ48120QNA08 Phone 1-888-567-9596 Doc.# 005-2QN412D C



Quarter Brick

48V_{in} 12V_{out} 8.3A





Shown Actual Size

- 1) All dimensions in inches (mm)
- 2) Pins 1-3, 5-7 are 0.040" (1.02mm) dia. with 0.080" (2.03mm) dia. standoff shoulders.
- 3) Pins 4 and 8 are 0.062" (1.57 mm) dia. with 0.100" (2.54mm) dia. standoff shoulders.
- 4) All pins are Brass with Tin/Lead plating over Nickel
- 5) Tolerances: x.xx in. +/-0.02 in. (0.5mm) x.xxx in. +/-0.010 in. (0.25mm)
- 6) Weight: 1.20 oz. (34 g)
- 7) Workmanship: Meets or exceeds IPC-A-610B Class II

Pin No.	Name	Function				
1	Vin(+)	Positive input voltage (35V - 75V)				
2	ON/OFF	TTL input to turn converter on and off, referenced to Vin(-), with internal pull up.				
3	Vin(-)	Negative input voltage				
4	Vout(-)	Negative output voltage				
5	SENSE(-)	Negative remote sense ¹				
6	TRIM	Output voltage trim ²				
7	SENSE(+)	Positive remote sense ³				
8	Vout(+)	Positive output voltage				

ABSOLUTE MAXIMUM RATINGS

Input Voltage:

Non-Operating: 100V continuous Operating: 80V continuous

100V 10µs transients, 1% duty cycyle

Input/Output Isolation Voltage: 2000V Storage Temperature: -55°C to +125°C Operating Temperature: -40°C to +115°C Voltage at ON/OFF input pin: +18V / -2V

Notes:

- 1. Pin 5 must be connected to Vout(-) at load.
- 2. Leave Pin 6 open for nominal output voltage.
- 3. Pin 7 must be connected to Vout(+) at load.

OPTIONS

The PQ48120QNA08 comes in two versions that differ by the sense of the logic used for the ON/OFF control signal. The PQ48120QNA08P version uses positive logic; meaning that the converter is on when the ON/OFF signal (Pin 2) is high. The PQ48120QNA08N version uses negative logic; the converter is on when the ON/OFF signal is low. Logic input is TTL compatible with an internal pull up.

Patents: SynQor is protected under various patents, including but not limited to U.S. Patent # 5,999,417.

SAFETY

The PQ48120QNA08 series of converters are UL 1950 recognized (US & Canada) with basic insulation rating and TUV certified to EN60950 requirements.

The converters also meet 72/23/EEC and 93/68/EEC directives as well as 94V-0 flammability requirements for board and plastic components.

An external input fuse must always be used to meet these safety requirements.



Quarter Brick

48V_{in} 12V_{out} 8.3A

PQ48120QNA08 ELECTRICAL CHARACTERISTICS

 $(T_A=25^{\circ}\text{C}, \text{ airflow rate}=300 \text{ LFM}, V_{in}=48\text{Vdc unless otherwise noted}; \text{ full operating temperature range is -40°C to +115°C ambient temperature with appropriate power derating.)}$

PARAMETER	NOTES and CONDITIONS	PQ48120QNA08			
		Min.	Тур.	Max.	Units
INPUT CHARACTERISTICS					
Operating Input Voltage Range		35	48	75	V
Input Under-Voltage Lockout					
Turn-On Voltage Threshold		32	33	34	V
Turn-Off Voltage Threshold		28.5	29.5	30.5	V
Lockout Hysteresis Voltage		2.5	3.5	4.5	V
Maximum Input Current	100% Load, 35Vin			3.5	A
No-Load Input Current			80	120	mA
Off Converter Input Current			1.6	3	mΑ
Inrush Current Transient Rating			.01		A ² s
Input Reflected-Ripple Current	P-P thru 10µH inductor; Figures 13 & 15		6		mA
OUTPUT CHARACTERISTICS					
Output Voltage Set Point		11.82	12.00	12.18	V
Output Voltage Regulation			_	2.1	ļ ,,
Over Load			±5_	±24	mV
Over Line			±5	±24	mV
Over Temperature	1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1	11.50	±90	±240	mV
Total Output Voltage Range	over sample load, line and temperature	11.58		12.42	V
Output Voltage Ripple and Noise	20MHz bandwidth; Figures 13 & 16		7.	1.50	
Peak-to-Peak	Full Load, 1µF ceramic, 10µF tantalum		75	150	mV
RMS	Full Load, 1 pF ceramic, 10 pF tantalum	^	15	30	mV
Operating Output Current Range	0	0	0 /	8.33	A
Output DC Current-Limit Inception	Output Voltage 10% Low; Figure 17	9	9.6	11	À
Short-Circuit Protection (redundant shutdown)		16	18	20	Α
DYNAMIC CHARACTERISTICS	100 H F: 00		<i>E.E.</i>		I D
Input Voltage Ripple Rejection	120 Hz; Figure 20		55		dB
Output Voltage Current Transient	100μF load cap, 5Å/μs; Figure 12		500		<u>بر</u>
Positive Step Change in Output Current	50% lo to 75% lo		500		mV
Negative Step Change in Output Current	75% lo to 50% lo		500 800		mV
Settling Time to 1% Turn-On Transient			800		μs
Turn-On Time	Figures 9 & 10	4	8	12	me
Start-Up Inhibit Period	-40°C to +125°C; Figure F	180	200	215	ms
Maximum Output Capacitance	Full load; 5% overshoot of Vout at startup	100	200	2,000	ms µF
EFFICIENCY	1 oil load, 3% overshool of vool at sidilop			2,000	μι
100% Load	Figure 1		88.0		%
50% Load	rigule i		88.4		% %
TEMPERATURE LIMITS FOR POWER DERATING CURVES			00.4		/0
	Package rated to 150°C			125	°C
Semiconductor Junction Temperature Board Temperature	Board rated to 165°C			125	000
Transformer Temperature				125	ပို့
Transformer Temperature	Figures 5 & 7			123	
ISOLATION CHARACTERISTICS Isolation Voltage	1	2000			V
Isolation Voltage Isolation Resistance		10			MΩ
Isolation Resistance		10	470		pF
			4/0		LΡ
FEATURE CHARACTERISTICS Switching Frequency		195	215	235	kHz
ON/OFF Control (Option P)		175	213	233	KIIZ
Off-State Voltage		-2		0.8	V
On-State Voltage		2.4		18	V
ON/OFF Control (Option N)		2.4		10	, v
Off-State Voltage		2.4		18	V
On-State Voltage		-2		0.8	ľ
ON/OFF Control (Either Option)		-2		0.0	\
Pull-Up Voltage	See circuit diagram; Figure E		Vin/6.5	9.2	V
Pull-Up Resistance	Pull up to Vin/6		40	7.2	kΩ
Output Voltage Trim Range	Across Pins 8 & 4; Figures B & C	-20	40	+10	%
Output Voltage Remote Sense Range	Across Pins 7 & 5	-20		+10	%
	ACIOSS I IIIS / CC J				/0
Output Over-Voltage Protection	Over full temp range; % of nominal Vout	11 <i>7</i>	122	127	% °C

Specifications subject to change without notice.



Quarter Brick

48V_{in} 12V_{out} 8.3A

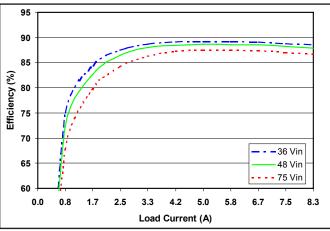


Figure 1: Efficiency vs. load current for minimum, nominal, and maximum input voltage at 25°C.

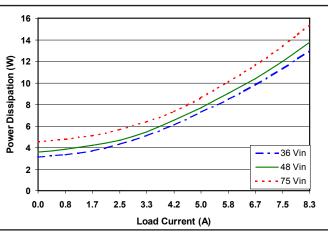


Figure 3: Power dissipation vs. load current for minimum, nominal, and maximum input voltage at 25°C.

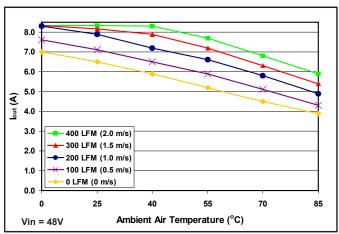


Figure 5: Maximum output power-derating curves vs. ambient air temperature for airflow rates of 0 LFM through 400 LFM with air flowing across the converter from pin 1 to pin 3 (nominal input voltage).

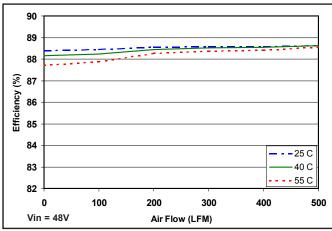


Figure 2: Efficiency at 60% rated power vs. airflow rate for ambient air temperatures of 25°C, 40°C, and 55°C and nominal input voltage.

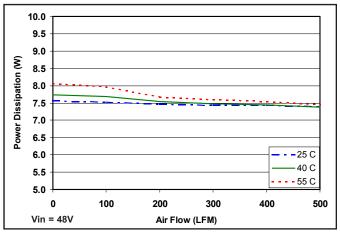


Figure 4: Power dissipation at 60% rated power vs. airflow rate for ambient air temperatures of 25°C, 40°C, and 55°C and nominal input voltage.

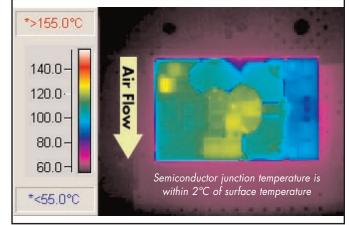


Figure 6: Thermal plot of converter at 6.6 amp load current with 55°C air flowing at the rate of 200 LFM. Air is flowing across the converter sideways from pin 1 to pin 3 (nominal input voltage).



Quarter Brick

48V_{in} 12V_{out} 8.3A

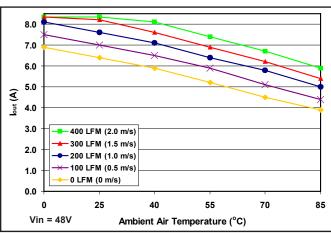


Figure 7: Maximum output power-derating curves vs. ambient air temperature for airflow rates of 0 LFM through 400 LFM with air flowing lengthwise from input to output (nominal input voltage).

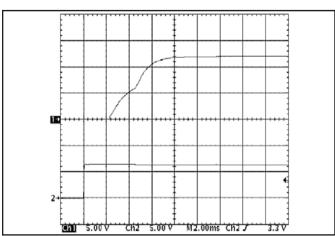


Figure 9: Turn-on transient at full rated load current (resistive load) (2 ms/div) Top Trace: Vout; 5V/div

Bottom Trace: ON/OFF input; 5V/div

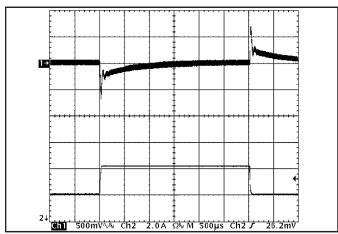


Figure 11: Output voltage response to step-change in load current (50%-75%-50% of Imax; $dI/dt = 0.1A/\mu s$). Load cap: $10\mu F$, $100 \text{ m}\Omega$ ESR tantalum capacitor and $1\mu F$ ceramic capacitor. Top trace: Vout (500mV/div), Bottom trace: Iout (2A/div).

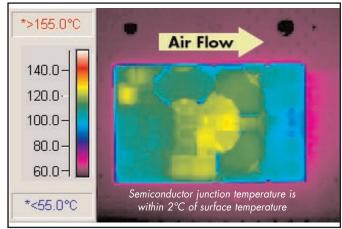


Figure 8: Thermal plot of converter at 6.4 amp load current with 55°C air flowing at the rate of 200 LFM. Air is flowing across the converter in the long direction from input to output (nominal input voltage).

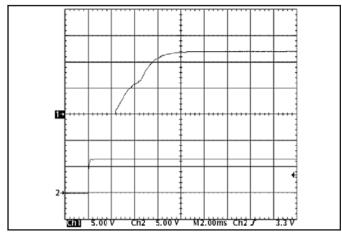


Figure 10: Turn-on transient at zero load current (2 ms/div).

Top Trace: Vout; 5V/div

Bottom Trace: ON/OFF input; 5V/div

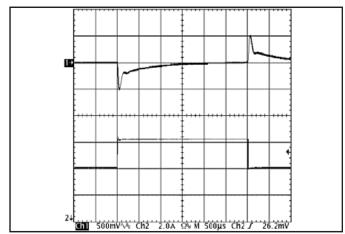


Figure 12: Output voltage response to step-change in load current (50%-75%-50% of Imax: $dI/dt = 5A/\mu s$). Load cap: $100\mu F$, $30 \, m\Omega$ ESR tantalum capacitor and $1\mu F$ ceramic capacitor. Top trace: Vout (500mV/div), Bottom trace: Iout (2A/div).



Quarter Brick

48V_{in} 12V_{out} 8.3A

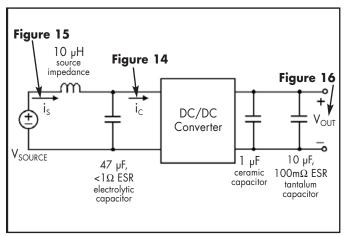


Figure 13: Test set-up diagram showing measurement points for Input Terminal Ripple Current (Figure 14), Input Reflected Ripple Current (Figure 15) and Output Voltage Ripple (Figure 16).

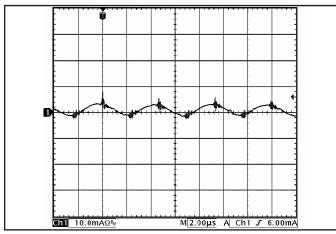


Figure 15: Input reflected ripple current, $i_{S'}$ through a 10 μ H source inductor at nominal input voltage and rated load current (10 mA/div). (See Figure 13)

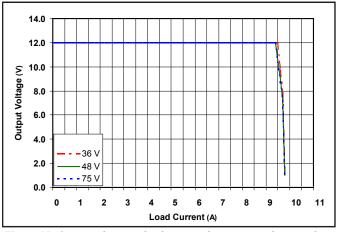


Figure 17: Output voltage vs. load current showing typical current limit curves and converter shutdown points.

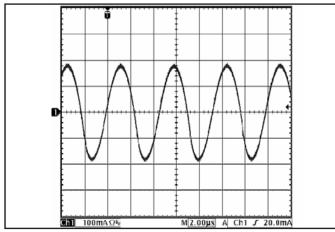


Figure 14: Input Terminal Ripple Current, i_c , at full rated output current and nominal input voltage with $10\mu H$ source impedance and $47\mu F$ electrolytic capacitor (100 mA/div). (See Figure 13)

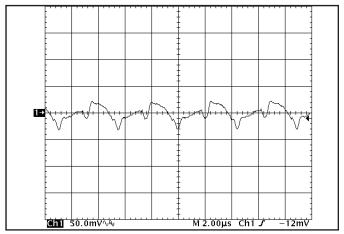


Figure 16: Output voltage ripple at nominal input voltage and rated load current (50 mV/div). Load capacitance: 1µF ceramic capacitor and 10µF tantalum capacitor. Bandwidth: 20 MHz. (See Figure 13)

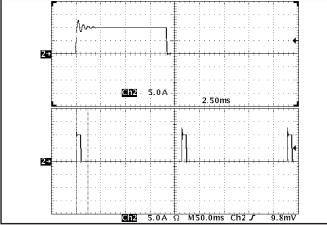


Figure 18: Load current (5A/div) as a function of time when the converter attempts to turn on into a 10 m Ω short circuit. Top trace is an expansion of the on-time portion of the bottom trace.



Quarter Brick

48V_{in} 12V_{out} 8.3A

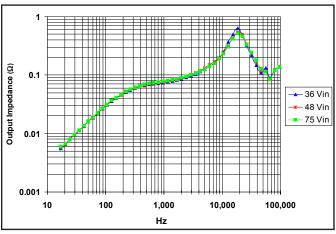


Figure 19: Output impedance ($Z_{out} = V_{out}/I_{out}$) for minimum, nominal, and maximum input voltage at full rated power.

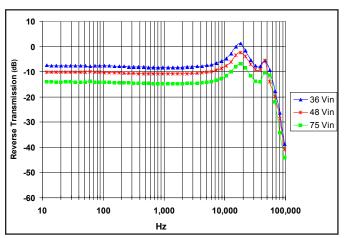


Figure 21: Reverse Transmission (RT = I_{in}/I_{out}) for minimum, nominal, and maximum input voltage at full rated power.

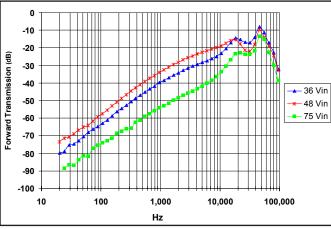


Figure 20: Forward Transmission ($FT = V_{out}/V_{in}$) for minimum, nominal, and maximum input voltage at full rated power.

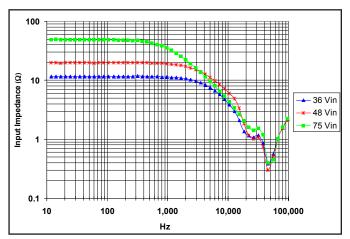


Figure 22: Input impedance ($Z_{in} = V_{in}/I_{in}$) for minimum, nominal, and maximum input voltage at full rated power.



BASIC OPERATION AND FEATURES

The PowerQor series converter uses a two-stage power circuit topology. The first stage is a buck-converter that keeps the output voltage constant over variations in line, load, and temperature. The second stage uses a transformer to provide the functions of input/output isolation and voltage stepdown to achieve the low output voltage required.

Both the first stage and the second stage switch at a fixed frequency for predictable EMI performance. Rectification of the transformer's output is accomplished with synchronous rectifiers. These devices, which are MOSFETs with a very low on-state resistance, dissipate far less energy than Schottky diodes used in conventional dc/dc converters. This is the primary reason that the *PowerQor* converter has such high efficiency—even at very low output voltages and very high output currents.

Dissipation throughout the converter is so low that the **PowerQor converter requires no heatsink** to deliver a greater level of power than can be delivered by a conventional, Schottky-diode-based dc/dc converter with a 0.5" high heatsink. At equivalent ambient air temperature, airflow rate, and output power level, the hottest semiconductor junction temperature and the hottest PCB temperature within the *PowerQor* converter are cooler than those found in conventional dc/dc converters with a 0.5" high heatsink attached.

Since a heatsink is not required, the *PowerQor* converter does not need a metal baseplate or potting material to help conduct the dissipated energy to the heatsink. The *PowerQor* converter can thus be built more simply using high yield surface mount techniques on a PCB substrate.

Unlike conventional dc/dc converters, which have critical thermal connections between the power components and the baseplate, and between the baseplate and the heatsink, the *PowerQor* converter has no explicit, failure-prone thermal connections.

Compared to a conventional Schottky-diode-based dc/dc converter with a 0.5" high heatsink, the *PowerQor* converter is more efficient and therefore it dissipates less than half the energy. Additionally, because the *PowerQor* converter is thinner (0.4" vs. 1.0"), the board-to-board pitch in a rack can be much smaller, and cooling airflow is less impeded by

Technical Specification

Quarter Brick

48V_{in} 12V_{out} 8.3A

the converter. Because the *PowerQor* converter is much lighter, vibration and shock-induced problems are greatly reduced. Moreover, due to the lack of failure-prone explicit thermal connections and the lack of potting material the *PowerQor* converter is more reliable than conventional dc/dc converters.

The *PowerQor* series converter uses the industry standard pin-out configuration used by other vendors of comparably sized and rated dc/dc converters. The unit is pin for pin compatible with the Lucent QW series.

The *Power*Qor converter has many standard control and protection features. All shutdown features are non-latching, meaning that the converter shuts off for 200ms before restarting. (see *Figure F*)

- An **ON/OFF** input permits the user to control when the converter is *on* and *off* in order to properly sequence different power supplies and to reduce power consumption during a standby condition.
- Remote sense inputs permit the user to maintain an accurate voltage at the load despite distribution voltage drops between the converter's output and the load.
- An output voltage trim input permits the user to trim the output voltage up or down to achieve a custom voltage level or to do voltage margining.
- An input under-voltage lockout avoids input system instability problems while the input voltage is rising.
- The output current limit protects both the converter and the board on which it is mounted against a short circuit condition.
- An output over-voltage limit circuit shuts the unit down if the output voltage at the output pins gets too high.
- A sensor located in a central spot of the PCB provides a
 PCB temperature limit. If, due to an abnormal condition, this spot gets too hot, the converter will turn off.
 Once the converter has cooled, it will automatically turn on again without the need to recycle the input power.

CONTROL PIN DESCRIPTIONS

Pin 2 (ON/OFF): The ON/OFF input, Pin 2, permits the user to control when the converter is *on* or *off*. This input is

SynCor

referenced to the return terminal of the 48V input bus. There are two versions of the *PowerQor* series converter that differ by the sense of the logic used for the ON/OFF input. In the PQxxyyyQNAzzPxx version, the ON/OFF input is active high (meaning that a high turns the converter *on*). In the PQxxyyyQNAzzNxx version, the ON/OFF signal is active low (meaning that a low turns the converter *on*). Figure A details five possible circuits for driving the ON/OFF pin.

Pins 7 and 5 (SENSE(\pm)): The SENSE(\pm) inputs correct for voltage drops along the conductors that connect the converter's output pins to the load.

Pin 7 should be connected to Vout(+) and Pin 5 should be connected to Vout(-) at the point on the board where regulation is desired. That is,

$$[Vout(+) - Vout(-)] - [SENSE(+) - SENSE(-)] \le 10\%Vout$$

Pins 7 and 5 must be connected for proper regulation of the output voltage. However, if these connections are not made, nothing catastrophic will happen to the converter under normal operating conditions—the converter will simply deliver an output voltage that is slightly higher than its specified value.

Note: the output over-voltage protection circuit senses the voltage across the output (pins 8 and 4) to determine when

Technical Specification

Quarter Brick

48V_{in} 12V_{out} 8.3A

it should trigger, not the voltage across the converter's sense pins (pins 7 and 5). Therefore, the resistive drop on the board should be small enough so that output OVP does not trigger, even during load transients.

Pin 6 (TRIM): The TRIM input permits the user to adjust the output voltage across the sense leads up or down. To lower the output voltage, the user should connect a resistor between Pin 6 and Pin 5, which is the SENSE(-) input. To raise the output voltage, the user should connect a resistor between Pin 6 and Pin 7, which is the SENSE(+) input.

A resistor connected between Pin 6 and Pin 5 will decrease the output voltage. For a desired decrease of Δ percent of the nominal output voltage, the value of this resistor should be

$$R_{trim\text{-}down} = \left(\frac{511}{\Delta\%}\right) - 10.22 \quad (k\Omega)$$
 where
$$\Delta = \left(\frac{V_{NOM} - V_{DES}}{V_{NOM}}\right) \times 100\%$$
 and
$$V_{NOM} = \text{Nominal Voltage}$$

$$V_{DES} = \text{Desired Voltage}$$

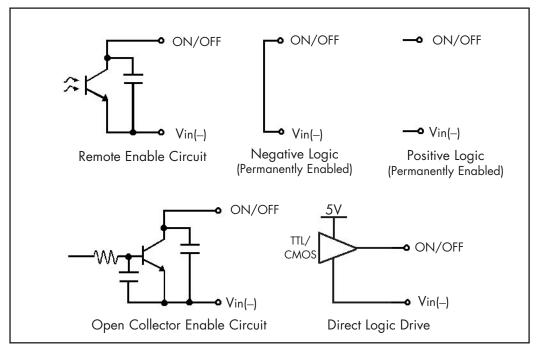


Figure A: Various circuits for driving the ON/OFF pin.

SynQor

Figure B graphs this relationship between $R_{trim\text{-}down}$ and Δ . The output voltage can be trimmed down as much as 20%.

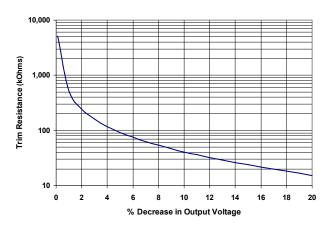


Figure B: Trim Down Graph for 12Vout

A resistor connected between Pin 6 and Pin 7 will increase the output voltage. For a desired increase of Δ percent of the nominal output voltage, the value of this resistor should be

$$R_{\text{trim-up}} = \left(\frac{5.11 V_{\text{OUT}} (100 + \Delta\%)}{1.225 \Delta\%} - \frac{511}{\Delta\%} - 10.22 \right) (k\Omega)$$

where

 V_{OUT} = Nominal Output Voltage

Figure C graphs this relationship between Rtrim-up and Δ . The output voltage can be trimmed up as much as 10%.

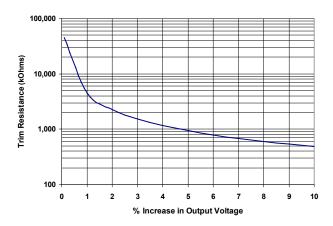


Figure C: Trim Up Graph for 12Vout

Technical Specification

Quarter Brick

48V_{in} 12V_{out} 8.3A

<u>Note</u>: the TRIM feature does not affect the voltage at which the output over-voltage protection circuit is triggered. Trimming the output voltage too high may cause the over-voltage protection circuit to engage, particularly during transients.

TOTAL DC VARIATION OF Vout: For the converter to meet its full specifications, the maximum variation of the dc value of Vout, due to both trimming and remote load voltage drops, should not be greater than +10%/-20%

PROTECTION FEATURES

Input Under-Voltage Lockout: The converter is designed to turn off when the input voltage is too low, helping avoid an input system instability problem, described in more detail below. The lockout circuitry is a comparator with dc hysteresis. When the input voltage is rising, it must exceed a typical value of 33V before the converter will turn on. Once the converter is on, the input voltage must fall below a typical value of 29.5V before the converter will turn off.

Output Current Limit: The current limit does not change appreciably as the output voltage drops. However, once the impedance of the short across the output is small enough to make the output voltage drop below approximately 60% of its nominal value, the converter turns off.

The converter then enters a mode where it repeatedly turns on and off at a 5 Hz (nominal) frequency with a 5% duty cycle until the short circuit condition is removed. This prevents excessive heating of the converter or the load board (see Figure 18).

Output Over-Voltage Limit: If the voltage across the output pins exceeds the O.V. threshold, the converter will immediately stop switching. This prevents damage to the load circuit due to 1) a sudden unloading of the converter, 2) a release of a short-circuit condition, or 3) a release of a current limit condition. Load capacitance determines exactly how high the output voltage will rise in response to these conditions. After 200 ms the converter will automatically restart.

Thermal Shutdown: The *PowerQor* series has a temperature sensor located such that it senses the average temperature of the converter. The thermal shutdown circuit is designed to turn the converter off when the temperature at



Quarter Brick

48V_{in} 12V_{out} 8.3A

the sensed location reaches 115°C. It will allow the converter to turn on again when the temperature of the sensed location falls below 110°C.

APPLICATION CONSIDERATIONS

Input System Instability: This condition can occur because a dc/dc converter appears incrementally as a negative resistance load. A detailed application note titled "Input System Instability" is available on the SynQor web site (www.synqor.com) which provides an understanding of why this instability arises, and shows the preferred solution for correcting it.

Application Circuits: Figure D below provides a typical circuit diagram which is useful when using input filtering and voltage trimming. Figure E is a detailed look of the internal ON/OFF circuitry that is shown in Figure A.

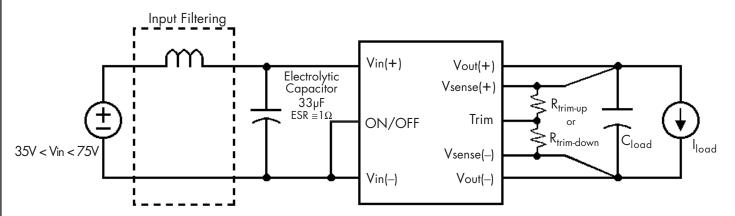


Figure D: Typical application circuit (negative logic unit, permanently enabled).

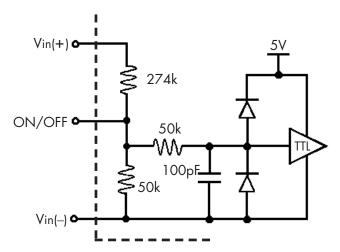


Figure E: Internal ON/OFF pin circuitry



STARTUP INHIBIT PERIOD

The Startup Inhibit Period ensures that the converter will remain off for at least 200ms when it is shut down for any reason. When an output short is present, this generates a 5Hz "hiccup mode," which prevents the converter from overheating. In all, there are seven ways that the converter can be shut down, initiating a Startup Inhibit Period:

- Input Under-Voltage Lockout
- Input Over-Voltage Shutdown (not present in Quarter-
- Output Over-Voltage Protection
- Over Temperature Shutdown
- Current Limit
- Short Circuit Protection
- Turned off by the ON/OFF input

Technical Specification

Quarter Brick

48V_{in} 12V_{out} 8.3A

Figure F shows three turn-on scenarios, where a Startup Inhibit Period is initiated at t_0 , t_1 , and t_2 :

Before time t₀, when the input voltage is below 34V (typ.), the unit is disabled by the Input Under-Voltage Lockout feature. When the input voltage rises above 34V, the Input Under-Voltage Lockout is released, and a Startup Inhibit Period is initiated. At the end of this delay, the ON/OFF pin is evaluated, and since it is active, the unit turns on.

At time t_1 , the unit is disabled by the ON/OFF pin, and it cannot be enabled again until the Startup Inhibit Period has elapsed.

When the ON/OFF pin goes high after t2, the Startup Inhibit Period has elapsed, and the output turns on within the 4ms (typ.) "Turn On Time."

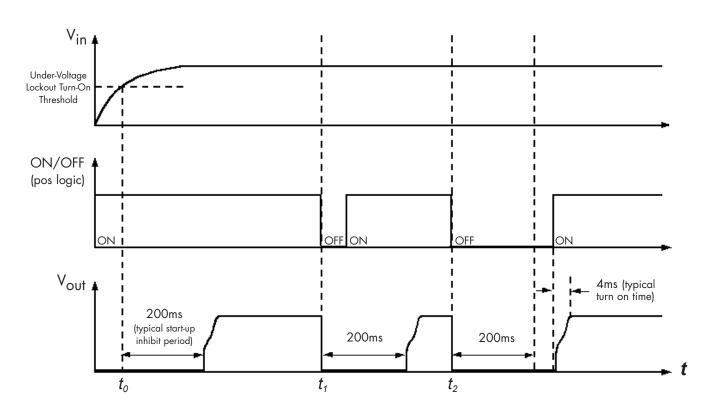


Figure F: Startup Inhibit Period (turn-on time not to scale)



Quarter Brick

48V_{in} 12V_{out} 8.3A

PART NUMBERING SYSTEM

The part numbering system for SynQor's *Power*Qor DC/DC converters has the following format:

Quarter-Brick Product Family and Part Numbering Scheme										
Product Family	Input Voltage	Output <u>Voltage</u>	Package Size	Performance <u>Series</u>	Thermal <u>Design</u>	"Rated" Output Current	Pos./Neg. Logic	Pin Length	<u>Features</u>	
PQ	48	033	Q	N	Α	25	N	N	S	
PQ - PowerQor	48 - (35v-75v)	015 - 1.5V 018 - 1.8V 020 - 2.0V 025 - 2.5V 033 - 3.3V 050 - 5.0V 060 - 6.0V 120 - 12V 150 - 15V	Q - Quarter Brick	N - Normal	A - Open Frame B - Baseplate	25 - 25 Amps 20 - 20 Amps 17 - 16.67 Amps 08 - 8.33 Amps 07 - 6.67 Amps	P - Positive N - Negative		S - Standard	
Base Part Number						Options				

Example part #: PQ48033QNA25NNS

This part number indicates a *PowerQor* converter with 48Vin, 3.3Vout, quarter-brick size, normal performance level, open air design, 25 amps output current, negative logic, 0.145" pins, and the standard feature set.

When ordering SynQor converters, please ensure that you use the complete 15 character part number.

Contact SynQor for further information:

<u>Phone</u>: 978-567-9596

<u>Toll Free</u>: 888-567-9596

<u>Fax</u>: 978-567-9599

E-mail: sales@synqor.com Web: www.synqor.com

Address: 188 Central Street Hudson, MA 01749 Warranty

SynQor offers a three (3) year limited warranty. Complete warranty information is listed on our web site or is available upon request from SynQor.

Information furnished by SynQor is believed to be accurate and reliable. However, no responsibility is assumed by SynQor for its use, nor for any infringements of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of SynQor.